

# MODELING AND DIRECT TORQUE CONTROL OF INDUCTION MOTOR BY USING FUZZY LOGIC CONTROL TECHNIQUE

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## Abstract

In this paper fuzzy rule based field oriented speed controlling of induction motor is proposed. The proposed speed controller estimate the suitable inputs in terms of logic variables to control the system. The speed of the motor is controlled by controlling the electromagnetic torque. To control the electromagnetic torque, the actual torque and change in torque is processed through fuzzy logic controller (FLC) which will produce the controlled electromagnetic torque in proportional to the speed of the motor. The proposed direct torque controller (DTC) improves the transient performance of induction motor with potential features. The entire proposed system is designed and tested under MATLAB/SIMULINK environment and simulation results also presented. And also the proposed fuzzy logic controller (FLC) performance is compared with the conventional Proportional-Integral (PI) controller performance to demonstrate the attractive performance of the proposed fuzzy logic controller.

**Key words:** Induction motor, DTC, fuzzy logic controller.

## 1. INTRODUCTION

Induction Machine is an important class of electric machines which finds wide applicability as a motor in industry. Generally, variable-speed drives for induction motors require both wide operating range of speed and fast torque response, regardless of load variations [17]. To improve the motor efficiency, the flux must be reduced by obtaining a balance between copper and iron losses [19]. The variations of electrical parameters, the inaccuracy of estimated fluxes will degrade the speed control performance [7]. The control of IM is complex due to its nonlinear nature, and the parameters change with operating conditions [4]. Speed estimation is an issue of particular interest with

induction motor drives, where the mechanical speed of the rotor is generally different from the speed of the revolving magnetic field [18].

Direct Torque Control (DTC) technique controls the torque and speed of the motor, which is directly based on the electromagnetic state of the motor [1]. The name direct torque control is derived from the fact that on the basis of the errors between the reference and the estimated values of torque and flux, it is possible to directly control the inverter states in order to reduce the torque and flux errors within the permissible limits [8]. The main advantages of DTC are robust and fast torque response, no requirements for coordinate transformation, requirements for PWM pulse generation and current regulators [20]. The DTC method is a simple and give fast transient response against the speed variations of the motor, hence most of the industrial drives are equipped with DTC [15]. The different induction motor control were in development currently. The present paper concerns with the implementation of Fuzzy rule based technique to improve the speed/torque characteristics of the induction motor. Techniques such as DTC with space vector modulation (SVPWM) technique, artificial intelligence techniques etc. [5] [9]

## 2. MATHEMATICAL MODEL OF DTC OF INDUCTION MOTOR

The mathematical model of the induction machine is used for analyzing the dynamic behavior of the motor. The change in the dynamic behavior of the motor affects the motor parameters such as speed, torque, resistance, flux etc. So, the dynamic model is required for analyzing the performance of the induction motor. The dynamic model of the induction motor is derived by transforming the three phase quantities into two phase direct and quadrature axes quantities. The mathematical equation of DTC of the induction motor is given below.

### 2.1 Voltage equations:

$$V_{s\alpha} = r_s i_{s\alpha} + P\psi_{s\alpha} \quad (1)$$

$$V_{s\beta} = r_s i_{s\beta} + P\psi_{s\beta} \quad (2)$$

$$V_{r\alpha} = r_r i_{r\alpha} + P\psi_{r\alpha} + \psi_{r\beta} \omega_r \quad (3)$$

$$V_{r\beta} = r_r i_{r\beta} + P\psi_{r\beta} - \psi_{r\alpha} \omega_r \quad (4)$$

$$\text{Where, } \psi_{s\alpha} = \psi_{r\alpha} = i_{r\alpha} L_r + i_{s\alpha} L_s \text{ and} \quad (4.a)$$

$$\psi_{s\beta} = \psi_{r\beta} = i_{r\beta} L_r + i_{s\beta} L_s. \quad (4.b)$$

These stator flux equations are rewritten as follows,

### 2.2 Stator Flux:

$$\psi_{s\alpha} = \int (V_{s\alpha} - r_s i_{s\alpha}) dt \quad (5)$$

$$\psi_{s\beta} = \int (V_{s\beta} - r_s i_{s\beta}) dt \quad (6)$$

The magnitude of the stator flux ( $\psi_s$ ) is the combination of  $\alpha$  - axis and  $\beta$  - axis flux which is expressed as follow,

$$|\psi_s| = \sqrt{\psi_{s\alpha}^2 + \psi_{s\beta}^2} \quad (7)$$

The magnetic track phase angle ( $\theta$ ) is expressed as follow,

$$\theta = \arctan\left(\frac{\psi_{s\alpha}}{\psi_{s\beta}}\right) \quad (8)$$

Where,  $V_{s\alpha}$ ,  $V_{s\beta}$ ,  $V_{r\alpha}$  and  $V_{r\beta}$  are the  $\alpha$  - axis and  $\beta$  - axis voltages of stator and rotor respectively. Then,  $i_{s\alpha}$ ,  $i_{s\beta}$ ,  $i_{r\alpha}$  and  $i_{r\beta}$  are the  $\alpha$  - axis and  $\beta$  - axis currents of stator and rotor respectively. The resistance and inductance of the stator and rotor winding are denoted as  $r_s$ ,  $r_r$ ,  $L_s$  and  $L_r$  respectively. The stator and the rotor flux of the motor are described as  $\psi_{s\alpha}$ ,  $\psi_{r\alpha}$ ,  $\psi_{s\beta}$  and  $\psi_{r\beta}$  respectively.

The electromagnetic torque of the motor is obtained from the inductance and the  $\alpha$  -  $\beta$  axis current. The  $\alpha$  -  $\beta$  axis current of the motor is obtained from the flux linkages that occurred in the stator and rotor of the motor. The expression of the electromagnetic torque that is produced by the motor is given as follows.

### 2.3 Electromagnetic Torque:

$$T_e = \frac{3}{2} \cdot \frac{P}{2} (\psi_{s\alpha} i_{s\beta} - \psi_{s\beta} i_{s\alpha}) \quad (9)$$

The actual torque developed by the motor in terms of rotor speed is illustrated as follow.

$$T = T_e - T_L = \frac{P}{2} \left( J \frac{d\omega_r}{dt} + B\omega_r \right) \quad (10)$$

Where,  $P$  is the number of poles,

$J$  is the moment of inertia of rotor,

$B$  is the damping coefficient,

$\omega_r$  is the rotor speed,

$T_L$  is the load torque,

$T_e$  is the electromagnetic torque.

### 2.4 Vector Diagram:

Considering the vector diagram shown in fig.1. The three phase currents  $i_a$ ,  $i_b$  and  $i_c$  are converted to a two-phase orthogonal system with axes  $\alpha$  and  $\beta$  by means of a technique called the Clarke Transformation. For simplicity, axis  $\alpha$  can be made equal to axis a. The two orthogonal currents  $i_\alpha$  and  $i_\beta$  are then transformed to a time-invariant, rotating orthogonal system represented by the field and torque components d and q of the equivalent rotor currents  $i_d$  and  $i_q$ . The  $\alpha/\beta$  coordinate frame is rotated counterclockwise to line up with the rotor flux axis  $\psi_r$ . The angle of rotation ( $\theta$ ) is determined with the help of the motor model. The vector diagram of flux linkages in the stator and rotor winding is illustrated as following.

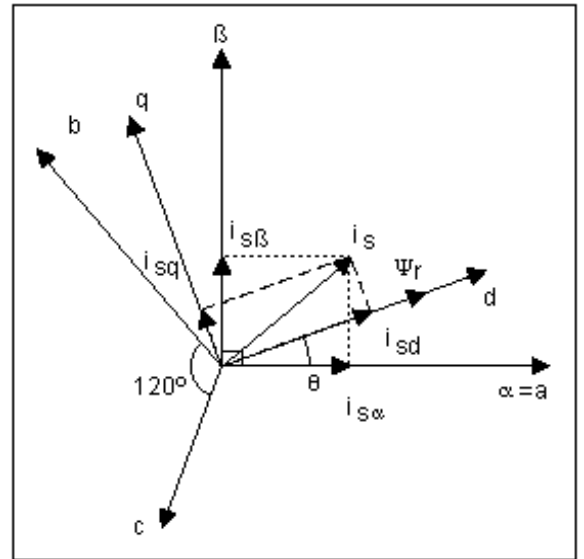


Fig. 1: Vector Diagram of DTC of Induction Motor

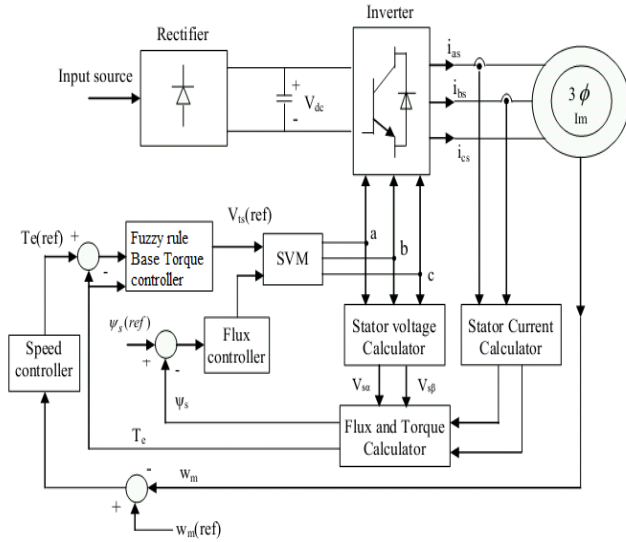


Fig. 2: Proposed fuzzy rule Based DTC System

In Figure 2, the control structure of proposed FLC is easily described. In the proposed model, the actual torque ( $T_e^*$ ) and change of torque ( $\Delta T_e^*$ ) are determined. The reference values of the torque are varied by the variation of load variation so, it's necessary to optimize reference torque. Then, the change of torque is determined from the torque deviated from the reference torque of the system. These actual torque and change torque is applied to FLC and the controlled output is determined. Based on the controlled output, the PWM control signal is generated for controlling the inverter gate. Here, the space vector PWM modulation technique is used for determining the gate control signal. The SVM based switching table is referred from [29] switching output, the IGBT gate is controlled. Then, the detailed description of fuzzy logic controller will be discussed in the preceding section.

### 3. FUZZY LOGIC CONTROLLERS

The conventional PI controllers are fixed-gain controllers i.e., the proportional and integral gains are constant. So, this type of controllers will not compensate properly, if the parameter changes and it does not adjust with changes in environment. The response of the PI-controller is too slow due to its sluggish response to relatively prompt variations in the state which will result more settling time. In addition to finding the gain constants related to system is very

difficult. Therefore the fuzzy control algorithm is capable of improving the system performance as compared with the classical methods. The two FLC input variables are the electromagnetic torque (T) and change of electromagnetic torque ( $\Delta T$ ). The operation of a fuzzy logic controller depends on the shape of membership functions for rule base.

In this paper a fuzzy logic control scheme (Figure 2) is proposed for speed control of an induction motor. The fuzzy logic controller have advantages to be robust and relatively simple to design. The irrespective of system exact model we can able to design fuzzy logic controller. The fuzzy rule based approach is done through three stages. They are Fuzzification, inference engine and defuzzification.

#### 3.1. Fuzzification

The fuzzy membership function values are assigned to the linguistic variables using five fuzzy subsets. In this paper the electromagnetic torque (T) and change in electromagnetic torque ( $\Delta T$ ) are selected as input variables of the system. In this paper the fuzzy logic controller is designed in Sugeno model. The input membership functions are shown in fig.3 the gbellmf type of membership functions are selected to control the system. . The range of each membership function is taking by taking torque and change in torque as the reference.

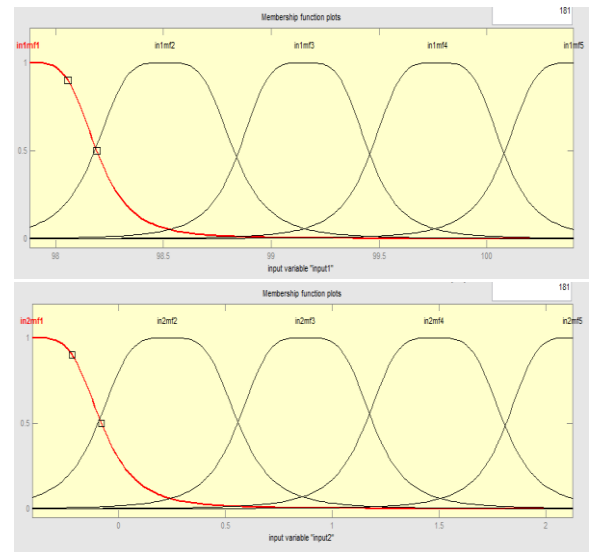


Fig.3 input membership functions.

### 3.2. Inference engine

This stage mainly consist of fuzzy rule base. In this rule base evaluation is done. In this stage first the inputs are fuzzified and this fuzzified inputs are fed through inference engine and rule base is applied. Then fuzzy output sets are identified.

### 3.3. Defuzzification

After identification of fuzzy output sets the defuzzification is needed because to obtain the controlled torque. The centroid defuzzification method is used defuzzification in the proposed technique. The output membership functions params values and the fuzzy rules are given in appendix-A. The controlled torque which is the output of the fuzzy logic controller is converted in to stator voltage. This is poses through the SVM, this will generates the firing pulses to the

inverter to control he speed of the induction motor which closely related to the reference speed.

## 4. SIMULATION

The proposed fuzzy rule based DTC technique was implemented in MATLAB/SIMULINK platform. Then, the speed control performance of proposed control technique was tested with induction motor of rating 2.2Kw/400V. The speed control performance of proposed control technique was compared with PI controller. From the proposed model, the speed, torque, current, flux and d-q axis voltage were analyzed. The Simulink model of proposed speed control system is illustrated in Fig.5. and the conventional PI controller model is shown in fig.4. The implementation parameters of proposed model is depicted in appendix-A.

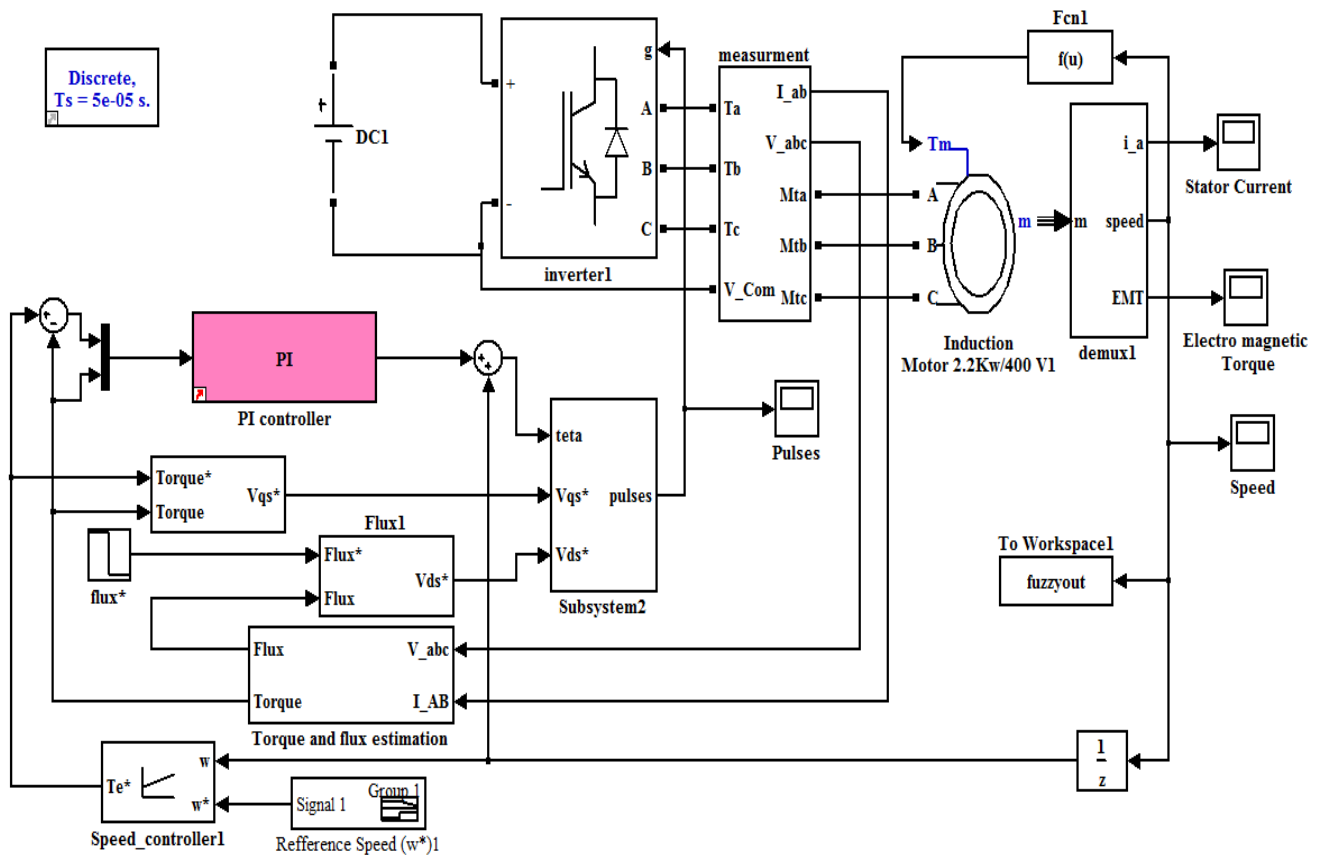


Fig. 4: Simulink Model of PI Speed control System.

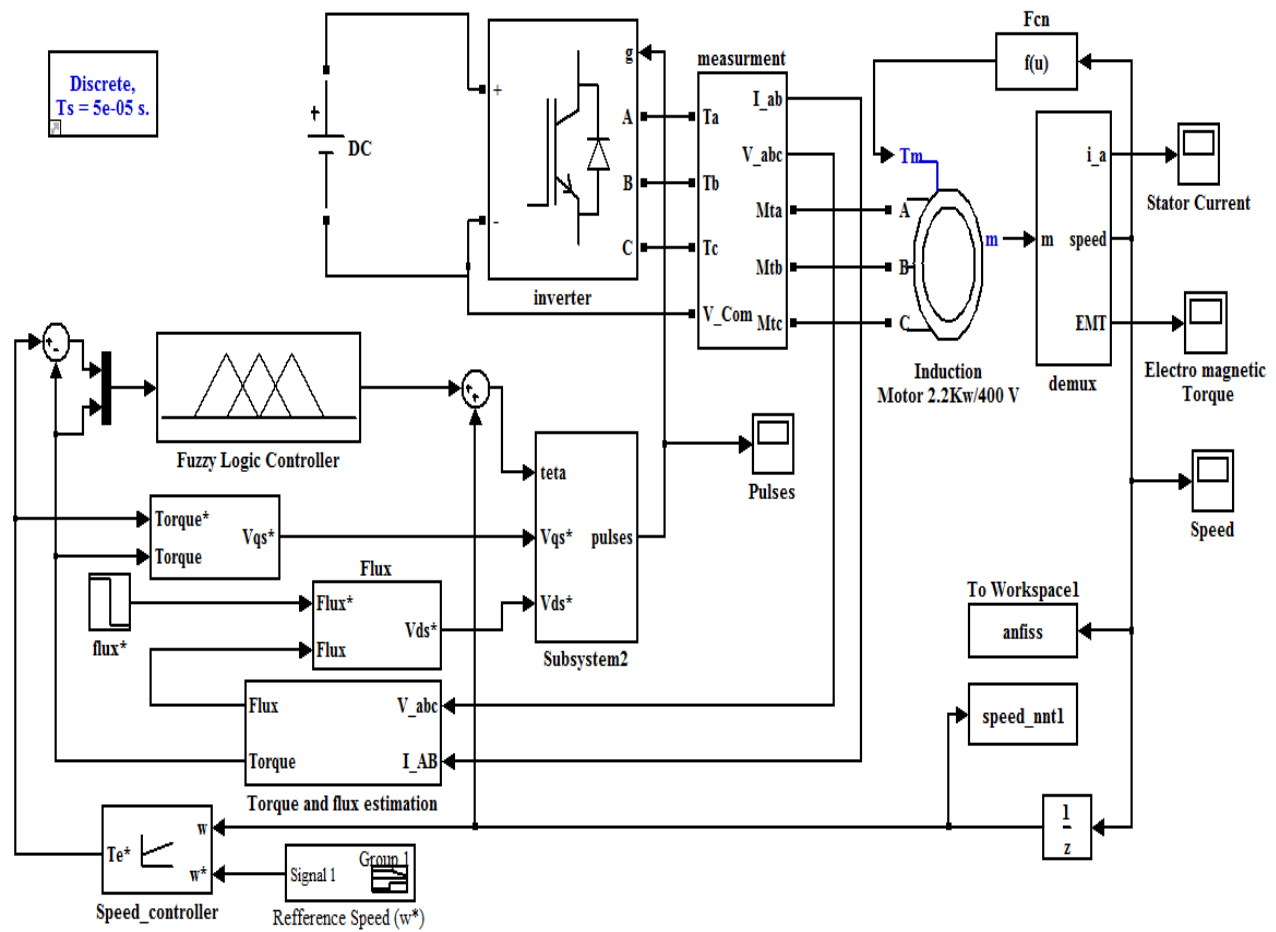


Fig. 5: Simulink Model of Fuzzy Logic Speed control System.

## 5. RESULTS AND DISCUSSION

From the above simulation results, the set point is described by the dynamic response of the IM when the speed set point is 80 rad/sec from rest at the duration of 0.5 sec. Then the speed set point is varied from 80 rad/sec to 60 rad/sec at duration of 0.5 sec. After that the speed set point is varied from 60 rad/sec to 100 rad/sec at duration of 1sec. The performance of current, voltage, flux, torque, speed and comparison of torque and speed are depicted from Fig. 6 to Fig. 17.

Figures (6), (7) and (8) shows the stator current, stator voltage and stator flux with PI controller respectively which are independent of set point.

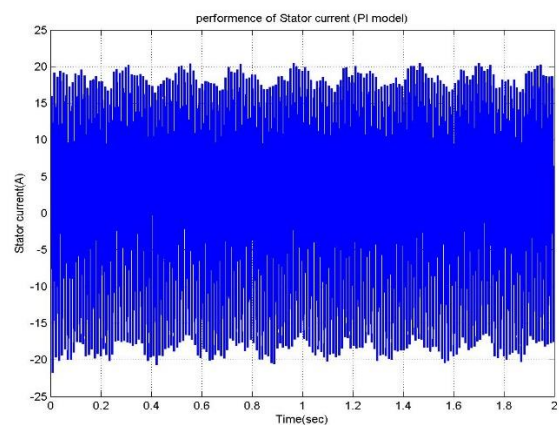


Fig. 6: Dynamics of Stator Current.

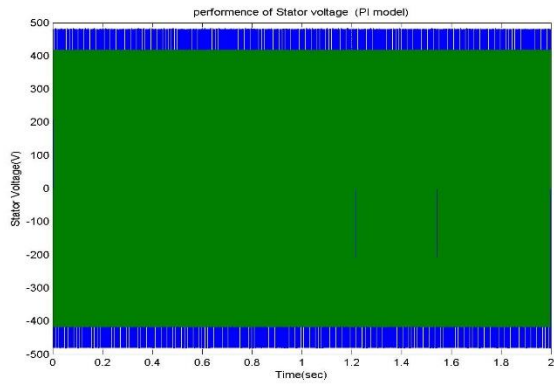


Fig. 7: Transformed d-q axis of Stator Voltage.

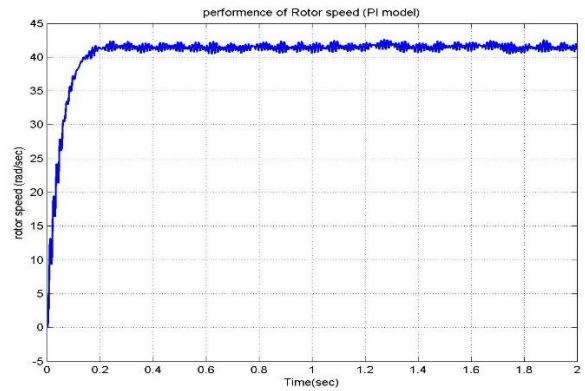


Fig. 10: Performance of Speed.

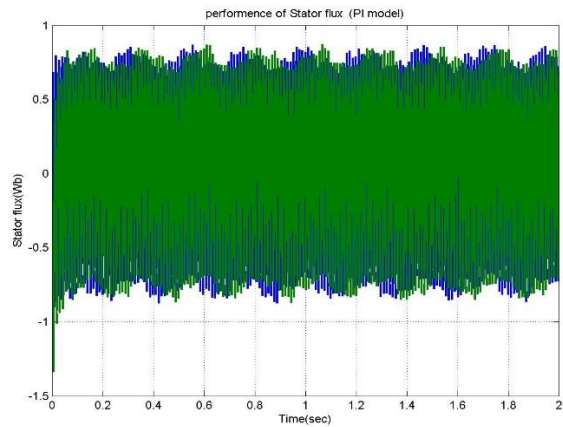


Fig. 8: Dynamics of d-q axis Stator Flux

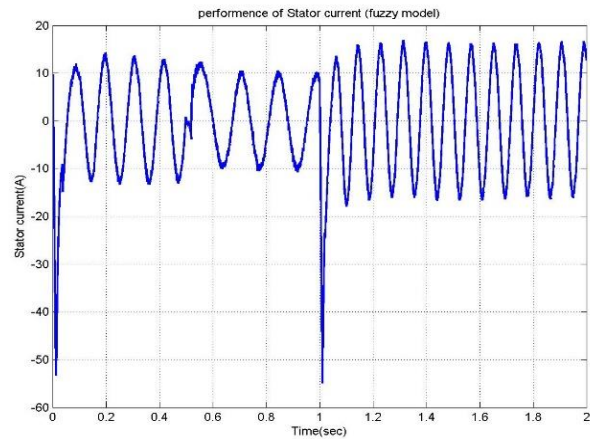


Fig. 11: Dynamics of Stator Current.

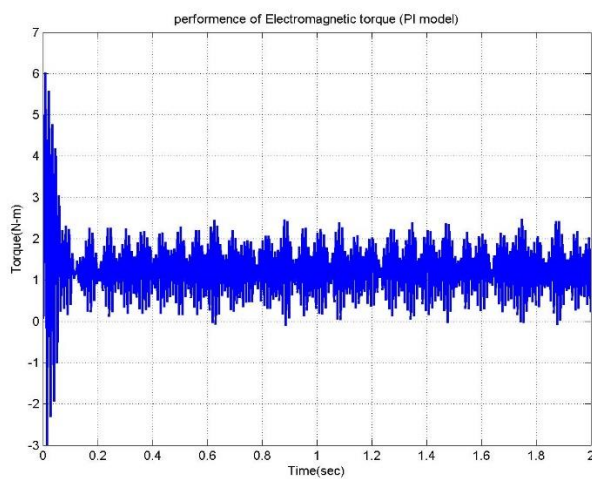


Fig. 9: Variations of Electromagnetic Torque.

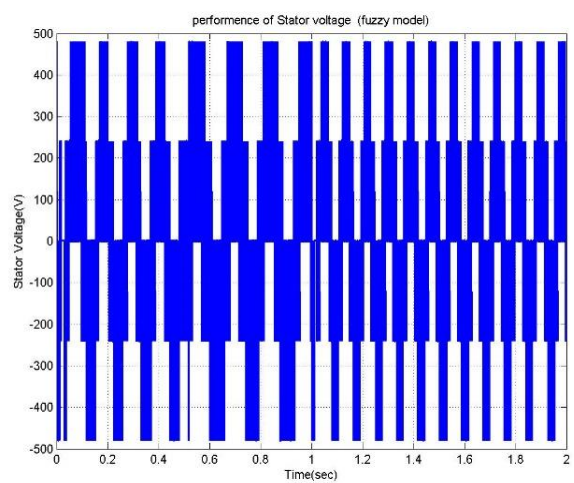


Fig. 12: Transformed d-q axis of Stator Voltage.

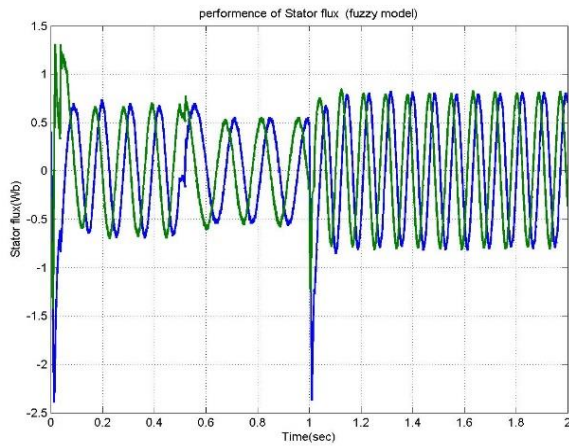


Fig. 13: Dynamics of d-q axis Stator Flux.

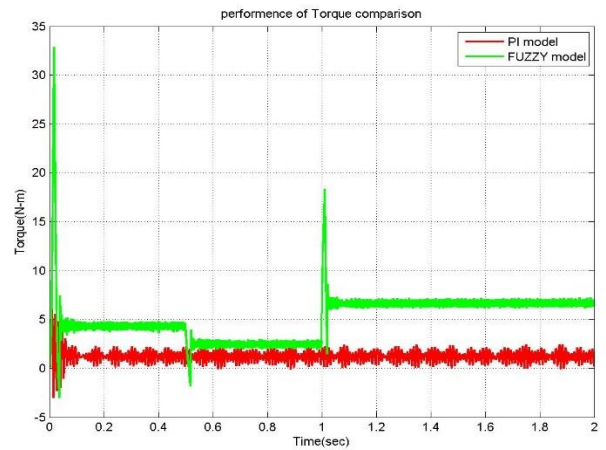


Fig. 16: Comparison Performance of Electromagnetic Torque

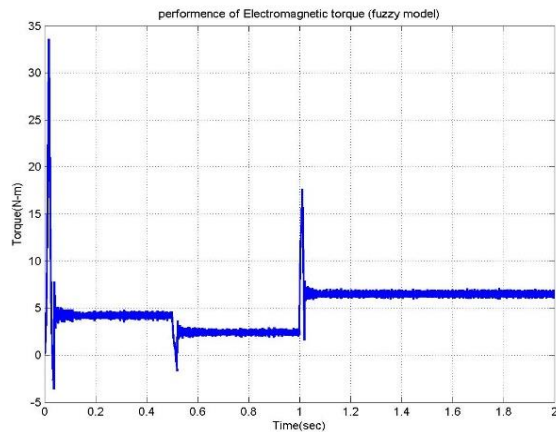


Fig. 14: Variations of Electromagnetic Torque

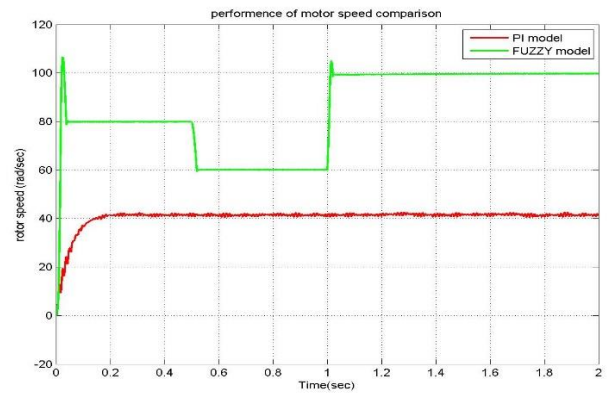


Fig. 17: Comparison Performance of Rotor Speed.

In figure 17, the comparison performance of the proposed Fuzzy speed controller and PI speed controller are described. From the comparative analysis, the proposed Fuzzy speed controller provides smooth speed control performance with less settling time, oscillations and steady state error when compared with the PI controller. The PI controller does not reach the steady state set point. Hence, the proposed FUZZY speed controller is better than PI speed controller.

## 6. CONCLUSIONS

In this paper fuzzy rule based field oriented speed controlling of induction motor is proposed. To evaluate the performance of the proposed technique was tested on 2.2Kw/400V induction motor. The mathematical

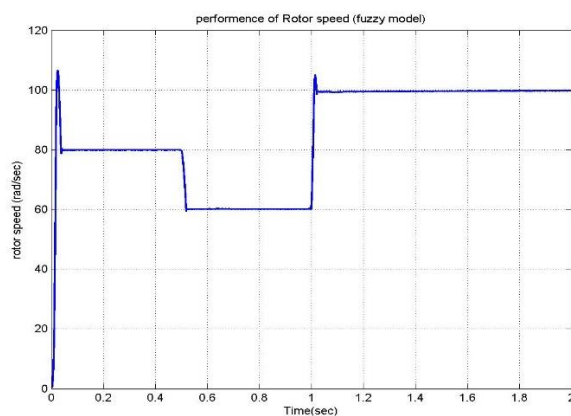


Fig.15: Performance of Speed.

model of the induction motor was analyzed in terms of stator voltage and torque equation. The actual torque and the error in torque was possess through a Fuzzy logic controller this estimates the controlled electromagnetic torque, finally this signal is applied to SVM by converting it into the stator voltage. The SVM generates the control signal to control the speed of the motor which is closely related to the reference speed. The performance speed and torque of the motor are analyzed and this performance characteristics are compared with PI controller. From the comparative analysis it is concluded that the proposed fuzzy logic control technique is better than PI control technique.

#### APPENDIX-A

Table I: Implementation parameters.

Parameters	Values
Nominal power	2200 watts
Line to line voltage	400 volts
Frequency	50 Hz
Stator resistance	3.67 ohm
Stator inductance	0.0269 H
Rotor resistance	2.1 ohm
Rotor inductance	0.0269 H
Mutual inductance	0.0324 H
Inertia	0.0155 kg.m <sup>2</sup>
Friction	0.0025 N.m.s
Number of poles	2

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